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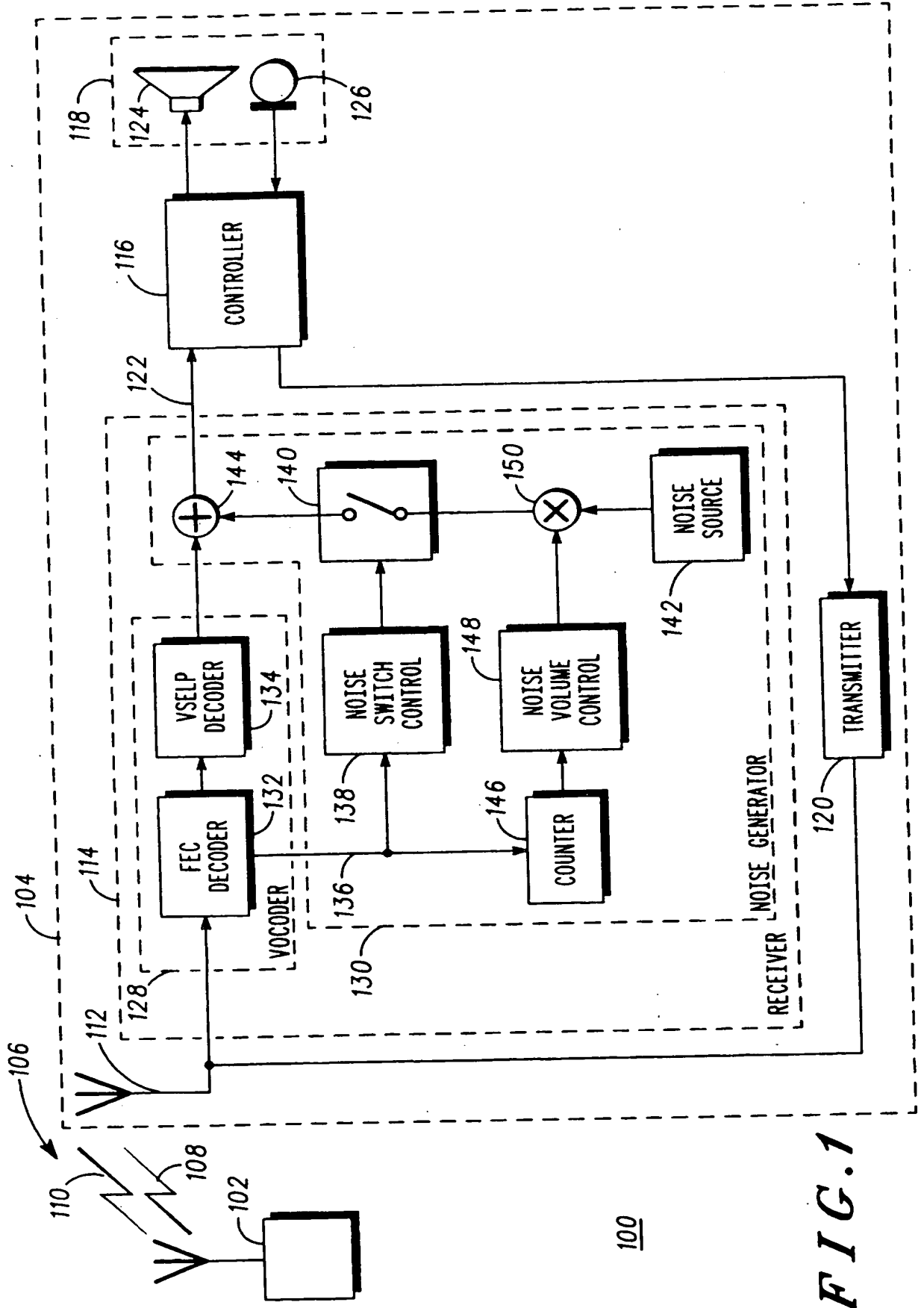


FIG. 1

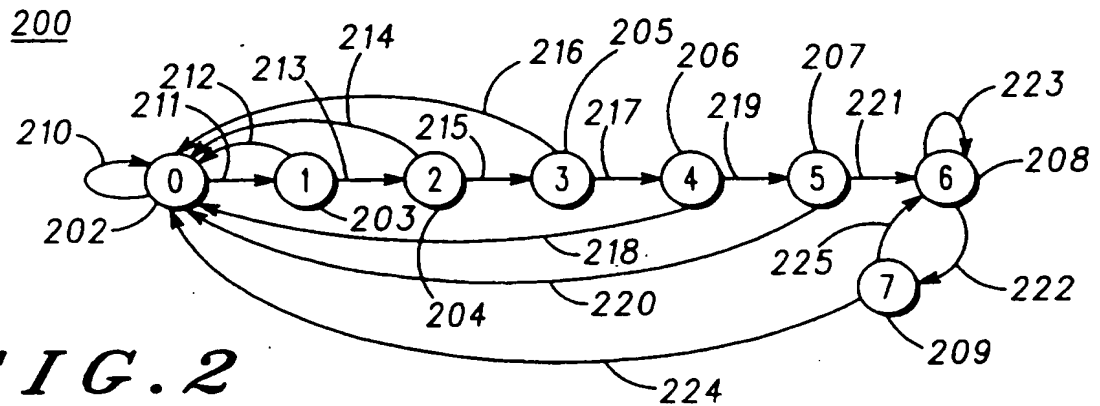


FIG. 2
—PRIOR. ART—

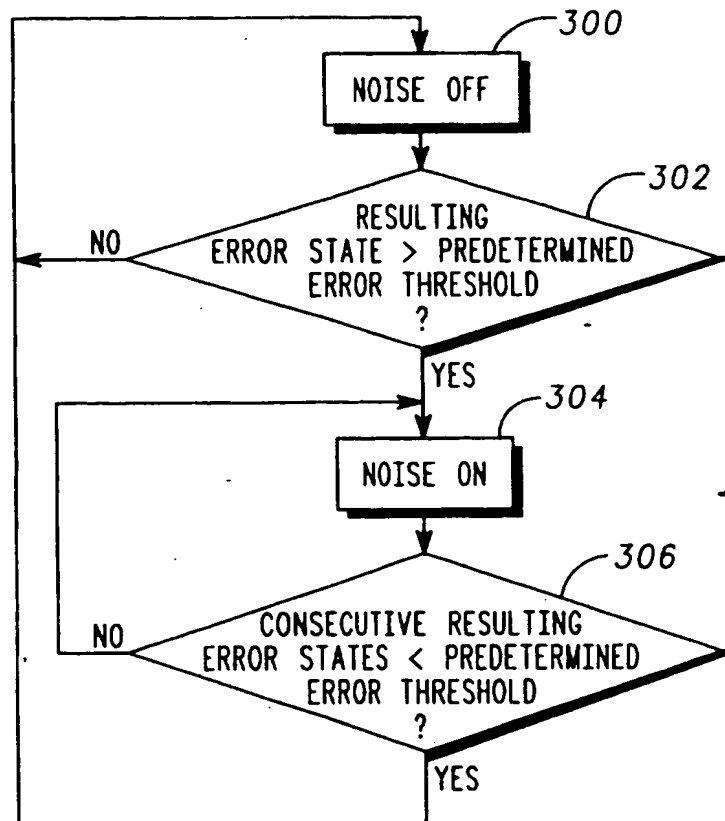


FIG. 3

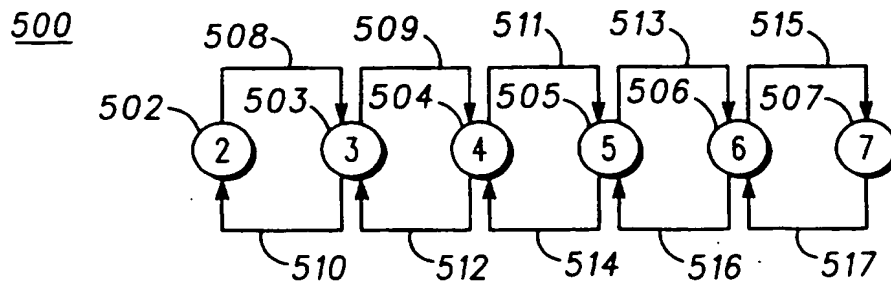
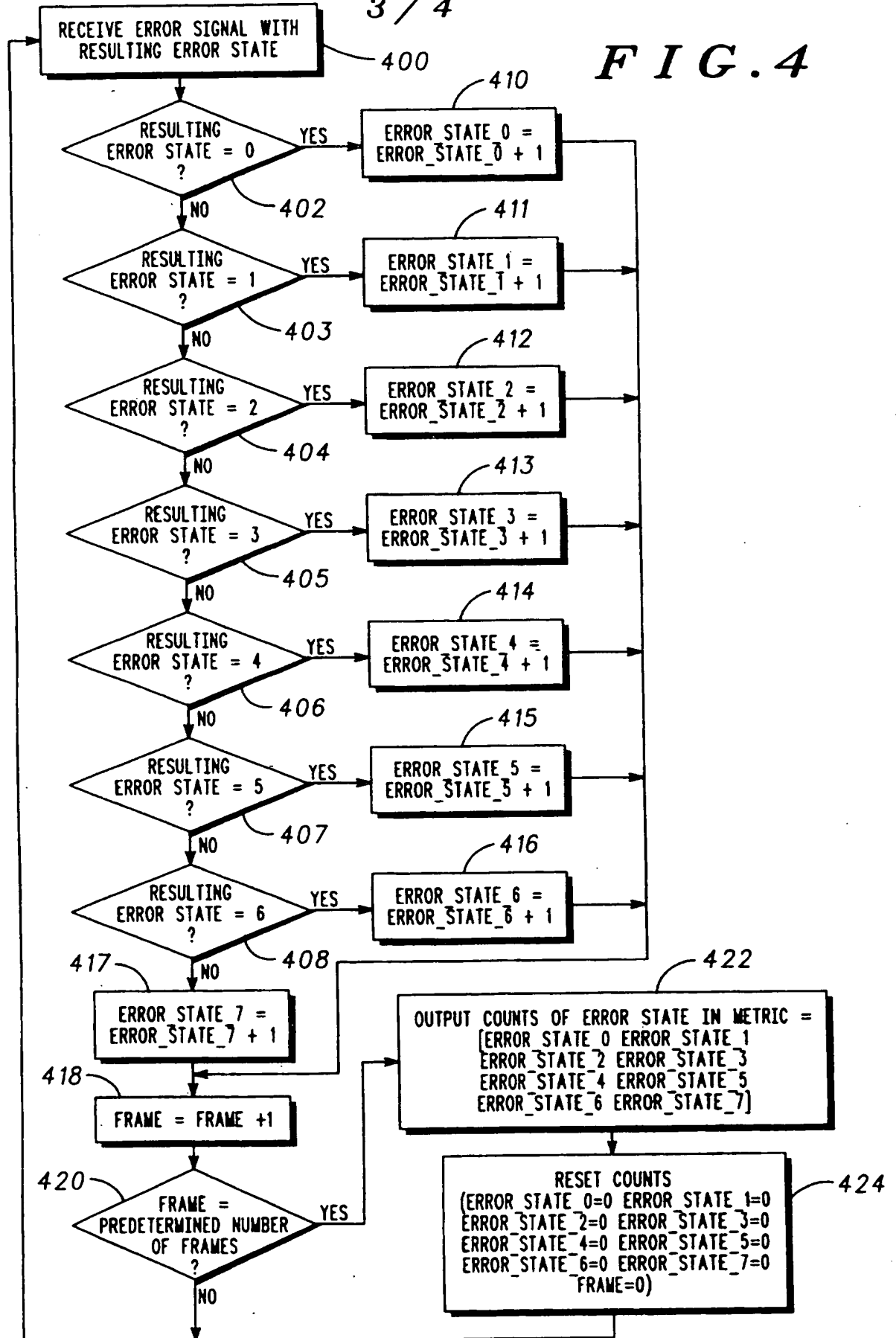
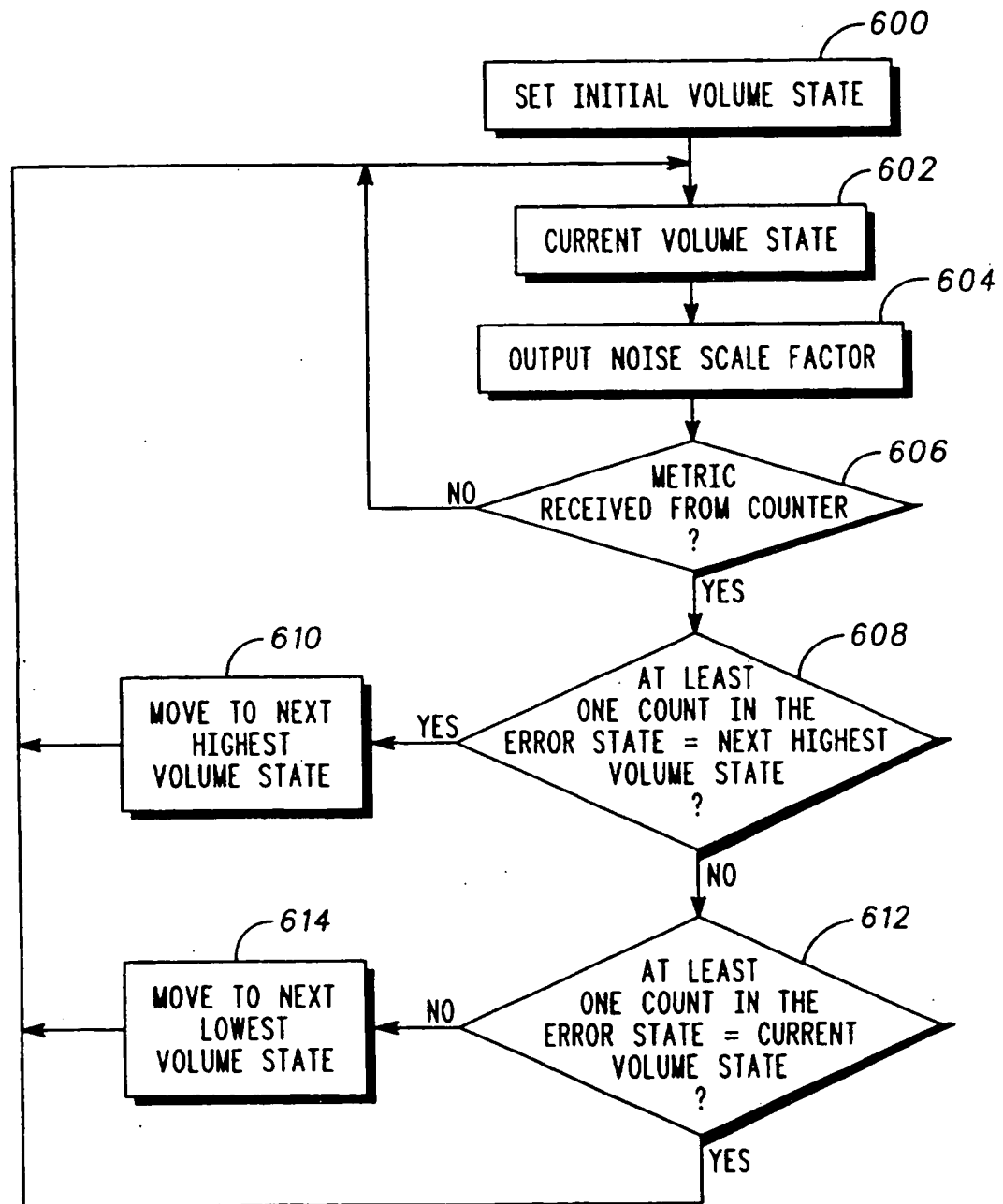


FIG. 5



*FIG. 6*

5

APPARATUS AND METHOD FOR GENERATING NOISE IN A DIGITAL RECEIVER

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Field of the Invention

The present invention relates generally to digital receivers and,
15 more specifically, to generating noise in a digital receiver.

Background of the Invention

20 Digital cellular telephones permit telephonic communication between users thereof. Digital cellular telephones employ a digital receiver that receives radio frequency (RF) signals and converts the RF signals to extract a data signal. The data signal is supplied to a vocoder which decodes the data signal, generates a speech signal therefrom, and
25 outputs the speech signal for use by a user. RF signals may become corrupted prior to, and during, reception. Corrupted RF signals may result in a corrupted data signal and a severely degraded speech signal.

The digital data signal extracted by the digital receiver, by its nature, lacks background noise unlike that present in traditional
30 analog receivers. As such, a traditional digital receiver does not produce noise in association with a corrupted speech signal. Traditional digital receivers alert the user to corrupted speech by letting the user hear distorted speech, issuing a special bad reception tone, or muting the speech. In these instances, the user can not tell if

telephonic communication has ended, the receiver has become inoperative, the antenna of the telephone needs reorienting, or the like.

5 A better approach to alerting the user of a digital receiver to corrupted speech is disclosed in U.S. Patent No. 5,327,457 by Leopold, assigned to Motorola Inc. In U.S. Patent No. 5,327,457, a background noise signal is generated based on a low signal level of received RF signals. The background noise signal is added to the speech signal to
10 improve user ambiance by making the digital receiver sound more like analog receivers with which users are more familiar. However, signal level is not always the best indicator that speech will be corrupted. For example, interference can cause corruption in receive RF signals without reducing signal level.

Therefore, it would be beneficial to carry the concept of U.S.
15 Patent No. 5,327,457 further by providing an improved apparatus and method for generating noise in a digital receiver.

Brief Description of the Drawings

20

FIG. 1 is a block diagram illustrating a digital communication system employing a digital communication device having a noise generator;

25 FIG. 2 is a known state diagram illustrating operation of a decoder block of the digital communication device of FIG. 1;

FIG. 3 is a flow chart illustrating operation of a noise switch control of the noise generator of FIG. 1;

FIG. 4 is a flow chart illustrating operation of a counter of the noise generator of FIG. 1;

30 FIG. 5 is a state diagram illustrating operation of a noise volume control of the noise generator of FIG. 1; and

FIG. 6 is a flow chart illustrating operation of the noise volume control of the noise generator of FIG. 1.

Detailed Description of the Preferred Embodiments

A digital receiver converts a received signal into a speech signal.
5 The circuitry generates an error signal indicative of a quality level of the received signal. The noise generator, responsive to receiving the error signal, selectively inserts noise into the speech signal. By basing noise insertion on the accuracy of the received signal, noise can be inserted into the speech signal only when the received signal is found
10 to be corrupted (i.e., the received signal is at a low quality level) such that the user will be subjected to degraded or muted speech.

FIG. 1 illustrates a digital communication system 100. The digital communication system 100 includes communication devices 102 and 104 that communicate over a communication link 106.
15 Communication devices 102 and 104 can be two-way radios, cellular telephones, cordless telephones, radios, base stations, radio transmitters, personal digital assistants, modems, land line telephones, or the like. The communication link 106 can be a wireless connection, a wireline connection such as a twisted wire pair or a coaxial cable, or
20 the like. In the illustrated embodiment, the digital communication system 100 is a digital cellular telephone system employing TDMA (Time Division Multiple Access), CDMA (Code Division Multiple Access), or the like. The illustrated communication device 102 is a cellular base station and the illustrated communication device 104 is a
25 digital cellular telephone compatible with communication device 102. In the illustrated embodiment, communication link 106 comprises a downlink radio frequency (RF) signal 108 and an uplink RF signal 110.

The communication device 104 includes an antenna 112, a receiver 114, a controller 116, a user interface 118, and a transmitter 120.
30 The antenna 112 couples the downlink RF signal 108 into the receiver 114 for demodulation and decoding into a received speech signal on line 122. The controller 116 couples the received speech signal to the user interface 118 where it is transduced by a speaker 124 into outgoing audible signals discernible by a user. Incoming audible signals

provided by a user are transduced via a microphone 126 of the user interface 118 and coupled to the transmitter 120 as a transmit speech signal via the user interface 118 and the controller 116. The transmitter 120 encodes and modulates the transmit speech signal for emission by the antenna 112 as the uplink RF signal 110.

The receiver 114 includes a vocoder 128 and a noise generator 130. The vocoder 128 and the noise generator 130 are implemented using a single digital signal processor (DSP) chip, such as the DSP56000 manufactured and sold by Motorola, Inc., or using one or more suitable microprocessors or microcontrollers. However, it will be recognized that the vocoder 128 and the noise generator 130 could instead be implemented wholly or partially as discrete components.

The vocoder 128 includes decoding circuitry comprising a forward error correction (FEC) decoder 132 and a vector sum excited linear prediction (VSELP) decoder 134. The FEC decoder 132 receives demodulated data of the downlink RF signal 108 at an input thereof and decodes the demodulated data into consecutive decoded frames of speech data according to a known decoding algorithm, such as a Viterbi algorithm. The FEC decoder 132 determines the quality level of each decoded frame of speech data by conducting a cyclic redundancy check (CRC) operation, such as that described in section 5.1.3.3 of Research and Development Center for Radio Systems Standard 27C (RCR STD-27C). The CRC operation detects bit errors in a portion of each decoded frame of speech data. Bit errors result from corruption of the downlink RF signal 108 that is not correctable by the decoding circuitry. Use of data from a corrupted downlink RF signal 108 causes severe degradation to the quality of the received speech signal. To prevent degradation to speech quality, the FEC decoder 132, responsive to the CRC, employs a bad frame masking state machine 200 (FIG. 2).

The bad frame masking state machine 200, also described in section 5.1.3.4 of RCR STD-27C, has error states 0-7, represented by blocks 202-209, respectively. The error states 0-7 represent levels of quality of the decoded frame of speech data. Transitions among the error states 0-7 are represented by the arrows 210-225 connecting the

blocks 202-209. A transition via one of the arrows 210-225 occurs for each decoded frame of speech data.

5 The bad frame masking state machine 200 begins at error state 0 (block 202). If no error is detected in the decoded frame of speech data, the bad frame masking state machine 200 remains at error state 0 (block 202), as represented by arrow 210. For each successive decoded frame of speech data in which an error is detected, the bad frame masking state machine 200 moves to the next higher numbered one of the error states 1 (block 203), 2 (block 204), 3 (block 205), 4 (block 206), 5 (block 207), and 6 (block 208), as represented by arrow 211, 213, 215, 217, 219, and 221, respectively. The error state number is thus the number of successive decoded frames in which an error is detected.

15 When a decoded frame is detected that does not include an error, the bad frame masking state machine 200 moves to a lower error state. The bad frame masking state machine 200 moves to error state 0 (block 202) from error state 1 (block 203), error state 2 (block 204), error state 3 (block 205), error state 4 (block 206), or error state 5 (block 207), as represented by arrow 212, 214, 216, 218, or 220, respectively, upon the occurrence of one decoded frame with no error detected.

20 The bad frame masking state machine 200 remains in error state 6 (block 208), represented by arrow 223, upon the occurrence of additional decoded frames including an error. Upon the occurrence of one decoded frame with no error, the bad frame masking state machine 200 moves from error state 6 (block 208) to error state 7 (block 209), as represented by arrow 222. Upon the occurrence of a decoded frame with an error while in error state 7 (block 209), the bad frame masking state machine 200 moves to error state 6 (block 208), as represented by arrow 225. If in error state 7 (block 209) and an errorless decoded frame occurs, the bad frame masking state machine 200 moves from error state 7 (block 209) to error state 0 (202), as represented by arrow 224.

30 At the conclusion of each transition, the FEC decoder 132 acts on the decoded frame of speech data according to a resulting one of the error states 0-7 (blocks 202-209). If the resulting error state is error state 0 (block 202), the quality of the decoded frame of speech data is at a

highest level and no action is taken by the FEC decoder 132 on the decoded frame of speech data. If the resulting error state is error state 1 (block 203) or error state 2 (block 204), the quality of the decoded frame of speech data is at a lower level and the FEC decoder 132 replaces the bits with errors of the decoded frame of speech data with corresponding bits from the most previous decoded frame of speech data that had a resulting error state of error state 0. If the resulting error state is error state 3 (block 205), error state 4 (block 206), or error state 5 (block 207), the quality of the decoded frame of speech data is at an even lower level and the FEC decoder 132 attenuates the decoded frame of speech data. If the resulting error state is error state 6 (block 208) or error state 7 (block 209), the quality of the decoded frame of speech data is at a lowest level and the FEC decoder 132 mutes the decoded frame of speech data. Next, the FEC decoder 132 outputs the decoded frame of speech data to the VSELP decoder 134 (FIG. 1). The VSELP decoder 134 generates the received speech signal on line 122 from the decoded frame of speech data. Muted speech data in the decoded frame of speech data creates audio voids in the received speech signal. A user listening at the speaker 124 is likely to believe that the voids in the speech indicate inoperability of the communication device 104.

To dispel these beliefs, at the end of each transition of the bad frame masking state machine 200, the FEC decoder 132 outputs an error signal on line 136. The error signal comprises the resulting error state for the decoded frame of speech data. The noise generator 130, responsive to the error signal, inserts noise into the received speech signal. The propagation delay through the noise generator 130 is approximately equal to that through the VSELP decoder 134. This ensures that the noise is inserted into the speech signal generated from the decoded frame of speech data containing the error.

The noise generator 130 comprises a noise switch control 138, a switch 140, a noise source 142, and an adder 144. The noise switch control 138 is coupled to the FEC decoder 132 via line 136 to receive the error signal. The noise switch control 138, responsive to the error signal, outputs a control signal. The switch 140 is coupled to the noise

switch control 138, the noise source 142, and the adder 144. The switch 140, in response to the control signal, selectively opens to disconnect the noise source 142 from the adder 144 or selectively closes to connect the noise source 142 to the adder 144. The noise source 142 outputs
5 noise samples to the switch 140. The noise samples are generated from real-time, pseudorandom, uniformly or Gaussian-distributed signals having a frequency spectrum such that the amplitude is flat within a voice range, such as 300 Hz to 3500 Hz. In the illustrated embodiment, the noise samples are white noise samples. White noise is preferred
10 because of its spectral flatness and smooth sounding characteristics. However, it will be recognized that other types of noise or alerts may alternatively be used. The adder 144 is further coupled to line 122. The adder 144 adds signals provided by the switch 140 to the received speech signal from the vocoder 128. Thus, the adder 144 adds the noise output
15 by the noise source 142 to the received speech signal when the switch is closed and, alternatively, the adder 144 does not add a signal to the received speech signal when the switch 140 is open.

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The specific operation of the noise switch control 138 is shown in FIG. 3. Initially, the noise switch control 138 outputs the control
20 signal to open the switch 140 and prevent noise from being added to the received speech signal (at block 300). The noise remains "off" (at block 300) until the noise switch control 138 detects a resulting error state of the error signal that exceeds a predetermined error threshold (at block 302). In the illustrated embodiment, the predetermined error
25 threshold is 3 and is exceeded by an error signal having a resulting error state of 4, 5, 6, or 7. Once an error signal of error state 4, 5, 6, or 7 is detected, the noise switch control 138 outputs the control signal to close the switch 140 and add noise to the received speech signal (at block 304). The noise remains "on" (at block 304) until the noise switch control 138
30 detects a predetermined number of consecutive error signals having a resulting error state that is below the predetermined error threshold (at block 306). In the illustrated embodiment, the noise switch control 138 must detect only resulting error states of 0, 1, or 2 in 250 consecutive error signals (each error signal representing a respective consecutive

the next error signal. If the predetermined number of decoded frames has occurred, the counter 146 outputs the counts of each of the error states for the previous predetermined number of decoded frames in the form of a metric (at block 422). In the illustrated embodiment, the metric comprises a one-dimensional array wherein the counts of the resulting error state 0 are in location 1, 1; the counts of resulting error state 1 are in location 1, 2; the counts of resulting error state 2 are in location 1, 3; the counts of resulting error state 3 are in location 1, 4; the counts of resulting error state 4 are in location 1, 5; the counts of resulting error state 5 are in location 1, 6; the counts of resulting error state 6 are in location 1, 7; and the counts of resulting error state 7 are in location 1, 8. For example, at the conclusion of 200 successive decoded frames, a metric of [160 20 6 6 4 2 1 1] would be output for the occurrence of 160 error signals with resulting error state 0, 20 error signals with resulting error state 1, 6 error signals with resulting error state 2, 6 error signals with resulting error state 3, 4 error signals with resulting error state 4, 2 error signals with resulting error state 5, 1 error signal with resulting error state 6, and 1 error signal with resulting error state 7. Once the metric is output, the counter 146 resets the error state and the frame counts (at block 424) and returns to block 400 to receive the next error signal associated with the next decoded frame.

The specific operation of the noise volume control 148 is shown in FIGs. 5 and 6. The noise volume control 148 employs a volume control state machine 500. The volume control state machine 500 comprises volume states 2-7, represented by blocks 502-507, respectively. The volume states numerically correspond to the error states in the metric output by the counter 146. In the illustrated embodiment, volume states 2-7 correspond to resulting error states 2-7, respectively. Each volume state corresponds to a noise scale factor output to the mixer 150. In the illustrated embodiment, the noise scale factors corresponding to volume states 2, 3, 4, 5, 6, and 7 (blocks 502, 503, 504, 505, 506, and 507) are approximately 0, 0.005, 0.008, 0.013, 0.018, and 0.023, respectively. Transitions among the volume states 2-7 (blocks

502-507) are represented by arrows 508-517. Transitions can occur after every reception of the metric from the counter 146.

5 The noise volume control 148 initially sets the volume control state machine 500 to a current volume state (at blocks 600 and 602). In the illustrated embodiment, the volume control state machine 500 is initially set to the volume state 2 (block 502). However, it will be recognized that the volume control state machine 500 could initially be set to any of the other volume states. The noise volume control 148 outputs a noise scale factor corresponding to the current volume state
10 (at block 604). The noise volume control 148 remains in the current volume state (i.e., volume state 2 (block 502)) until the metric is received from the counter 146 (at block 606).

15 The noise volume control 148 examines the metric to determine if the metric contains at least one count in an error state that is numerically equal to the next highest volume state (i.e., equal to the current volume state + 1) (at block 608). If so, the volume control state machine 500 moves from the current volume state to the next highest volume state (at block 610). The noise volume control 148 returns to block 602 where the next highest volume state becomes the current
20 volume state. For example, if the metric is [150 40 7 1 1 1 0 0] and the current volume state is volume state 2 (block 502), the presence of an error in error state 3 will cause the volume control state machine 500 to move to volume state 3 (block 503). In the illustrated embodiment, the possible transitions at block 610 are from volume
25 state 2 (block 502), 3 (block 503), 4 (block 504), 5 (block 505), or 6 (block 506) to volume state 3 (block 503), 4 (block 504), 5 (block 505), 6 (block 506), or 7 (block 507) via arrow 508, 509, 511, 513, or 515, respectively. Volume state 3 will be the current volume state when the noise volume control 148 returns to block 602.

30 If the metric does not contain at least one count in an error state that is numerically equal to the next highest volume state, the noise volume control 148 examines the metric to determine if the metric contains at least one count in an error state that is numerically equal to the current volume state (at block 612). If not, the volume control state

machine 500 moves from the current volume state to the next lowest volume state (at block 614). For example, if the metric is [150 40 5 5 0 0 0 0] and the current volume state is volume state 4 (block 504), the volume control state machine 500 moves to volume state 3 (block 503).

5 In the illustrated embodiment, the possible transitions at block 614 are from volume state 3 (block 503), 4 (block 504), 5 (block 505), 6 (block 506), or 7 (block 507) to volume state 2 (block 502), 3 (block 503), 4 (block 504), 5 (block 505), or 6 (block 506) via arrow 510, 512, 514, 516, or 517, respectively. The noise volume control 148 returns to block 602 where

10 the next lowest volume state becomes the current volume state.

If the metric does include at least one count in an error state that is numerically equal to the current volume state, the noise volume control 148 returns to block 602 where the current volume state is maintained. For example, if the current volume state is volume state 4

15 (block 504), such a metric would be [150 40 5 4 1 0 0 0].

To provide a better sounding noise that best mimics the continuous nature of noise in an analog radio, the noise volume control 148 (at block 604) outputs the current noise scale factor according to a ramping technique. The ramping technique includes

20 gradually increasing or decreasing the noise scale factor when moving between volume states. In the illustrated embodiment, the noise volume control 148 increments or decrements the noise scale factor over a period of 100 decoded frames or 2 secs. in steps defined by the difference between the previous noise scale factor and the current noise

25 scale factor divided by 100. For example, in response to transitioning from volume state 2 where the noise scale factor is 0 to volume state 3 where the noise scale factor is 0.005, the noise volume control 148 incrementally increases the noise scale factor 100 times, by 0.00005 $((0.005-0)/100)$ each time, over a time period of 2 secs. In response to

30 transitioning from volume state 4 where the noise scale factor is 0.008 to volume state 3 where the noise scale factor is 0.005, the noise volume control 148 incrementally decreases the noise scale factor 100 times, by 0.00003 $((0.008-0.005)/100)$ each time, over a time period of 2 secs.

Although based on the detection of bit errors, it will be recognized that noise insertion and volume control could alternatively be based on the detection of other vocoder artifacts that affect signal quality. The error signal can alternatively define a measure of phase jitter, timing jitter, eye opening jitter, constellation jitter, signal-to-noise ratio, or other vocoder artifact.

It will be recognized that the error signal need not be generated by the decoding circuitry or solely from the decoded input signal. For example, the error signal could alternatively be generated outside of the decoding circuitry and based on a comparison of a re-encoded decoded receive signal and the actual receive signal based as described in U.S. Patent No. 5,113,400 entitled "Error Detection System" by Gould et al., which was issued on May 12, 1992 and is assigned to Motorola, Inc.

Thus, it can be seen that noise can be inserted into a corrupted speech signal of a digital receiver to audibly indicate poor reception to the user. A noise generator adds noise to a speech signal based on the quality of the signal received by the digital receiver. In addition, the noise generator can gradually vary the volume of the noise based on further variations of signal quality. As the signal quality decreases, the volume level of the noise is increased. Alternatively, as the signal quality increases, the volume level of the noise is decreased. In one embodiment, the noise generator operates responsive to the decoder and accurately and efficiently inserts noise at an appropriate volume level directly into those frames of the speech signal determined to be corrupted by the decoder.

What is claimed is:

Claims

1. A digital receiver for converting a received signal into a speech signal, the digital receiver comprising:
 - 5 circuitry to generate an error signal indicative of a quality level of the received signal; and
 - a noise generator coupled to an output of the circuitry, the noise generator, selectively responsive to the error signal, to insert noise into the speech signal.
- 10 2. The digital receiver according to claim 1 wherein the circuitry to generate further includes a decoder, the decoder to decode the received signal and to detect errors therein, the decoder to generate the error signal based on the detected errors.
- 15 3. The digital receiver according to claim 1 wherein the noise generator further comprises:
 - a noise source to provide the noise, and
 - a switch coupled to the noise source and the output of the
 - 20 circuitry, the switch, responsive to the error signal, to selectively connect the noise source to the digital receiver at the speech signal.
- 25 4. The digital receiver according to claim 1 wherein the noise generator further comprises:
 - a noise source to provide the noise, and
 - a noise volume control coupled to the noise source, the noise
 - volume control, responsive to multiple occurrences of the error signal, to selectively vary a volume level of the noise, the volume level
 - 30 corresponding to the quality level of the error signal.

5. A method of generating noise in a digital receiver, the method comprising the steps of:
- converting a received signal into a speech signal;
 - providing an error signal indicative of a quality level of the received signal; and
 - selectively inserting, responsive to the error signal, noise into the speech signal.
6. The method according to claim 5 wherein the step of selectively inserting further comprises the substeps of:
- generating the noise;
 - detecting the error signal; and
 - selectively connecting the noise to the speech signal.
7. The method according to claim 6 wherein the step of selectively connecting further comprises the substeps of:
- connecting the noise to the speech signal on a first occurrence of the error signal at a low quality level; and
 - disconnecting the noise from the speech signal after a plurality of occurrences of the error signal at a high quality level.
8. The method according to claim 5 wherein the step of selectively inserting further comprises the substeps of:
- generating the noise;
 - detecting the error signal;
 - setting a volume level of the noise, the volume level corresponding to the quality level; and
 - selectively varying, responsive to multiple occurrences of the error signal, the volume level of the noise.

9. The method according to claim 8 wherein the step of selectively varying further comprises the substeps of:

counting a predetermined number of the error signal;

5 counting, during the step of counting a predetermined number of the error signal, each occurrence of the error signal at one of a high quality level and a low quality level; and

counting, during the step of counting a predetermined number of the error signal, each occurrence of the error signal at the other one of the high quality level and the low quality level.

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10. The method according to claim 9 wherein the step of selectively varying further comprises the substeps of:

increasing, responsive to the steps of counting, the volume level of the noise when any occurrences of the error signal indicate a quality

15 level lower than the volume level; and

decreasing, responsive to said steps of counting, the volume level of the noise when all occurrences of the error signal indicate a quality level higher than the volume level.



Application No: GB 9714440.6
Claims searched: 1-10

Examiner: D. Midgley
Date of search: 16 September 1997

Patents Act 1977
Search Report under Section 17

Databases searched:

UK Patent Office collections, including GB, EP, WO & US patent specifications, in:
UK Cl (Ed.O): H4L LFM H4P PEX
Int Cl (Ed.6): H04B 1/10 1/16 H04L 1/20
Other: ONLINE:WPI

Documents considered to be relevant:

Category	Identity of document and relevant passage	Relevant to claims
X,P	GB 2307623 A (NORTHERN) See, especially, page 2, lines 23-37	1-6,8 at least
X	EP 0464911 A2 (PHILIPS) See, especially, col. 3, line 38 to col. 4, line 9	1,5 at least

X	Document indicating lack of novelty or inventive step	A	Document indicating technological background and/or state of the art.
Y	Document indicating lack of inventive step if combined with one or more other documents of same category.	P	Document published on or after the declared priority date but before the filing date of this invention.
&	Member of the same patent family	E	Patent document published on or after, but with priority date earlier than, the filing date of this application.

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